

An Energy-Efficient 16-Channel Charge Trackable Inductor-Based Stimulation System Eojin Kim¹, Chul Kim^{*1}

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Abstract

These days, the concept of "electroceuticals" is emerging that maximizes the effectiveness of treatment by applying electrical stimulation only to the desired area but minimizes side effects. In order to realize electroceuticals, stimulators that can be inserted into the body must be developed. The requirements are such as enhancing power efficiency, securing safety, and effective electrical stimulation. Therefore, the stimulator must operate energy efficiently and must be able to track the amount of stimulation charges delivered to the target cell. However, there is a problem that the existing electrical stimulator does not satisfy the above requirements. Therefore, this study developed Charge Trackable Inductor-Based Stimulator (CTIBS), which operates in 80% of peak stimulator efficiency, and its charge error between anodic and cathodic stimulation to be less than 1.36%. In addition, the stimulation channels are expanded to 16-channels so that stimulation area and density can be freely controlled. The IC chip was designed in RF 180nm process, taking up the area of 5mm²

Design

Charge Trackable Inductor-Based Stimulator (CTIBS)



CTIBS consists of a stimulation capacitor, inductor, MOS switches. It is directly connected to 16 H-bridge stimulation channels with the electrode-tissue model (ETI). The MOS switches are controlled by commands transmitted from the digital circuit and the sensor block according to the four states of the system and the four phases of the stimulation state.

In CTIBS, the stimulation capacitor measures the amount of stimulation charge, enabling charge tracking and balance. The inductor returns the excess energy remaining in the stimulation circuit to the supply so that the stimulation circuit works in an energy-efficient way.

The Entire System Structure



The system is largely divided into a digital block, a biasing block, a sensor block, and a stimulation circuit.

These four blocks proceed with 4 states and 4 stimulation phases through interaction. The four states consist of an idle state (IDLE), an intermediate delay state (INTER), and cathodic/anodic stimulation state (CAT/AN_STIM). Each stimulation state consists of four phases, which grants energy-efficiency and charge tracking ability to the proposed stimulator. The four phases consist of stimulation phase (STIM), transfer phase (TRAN), replenishing phase (REP), and dump phase (DUMP).

Double Comparator Phase Control

Bad Good

As the STIM phase is repeated, the charge accumulates in the double layer capacitor of the electrode. This changes the amplitude of the stimulation current, thus the duration taken to bring/send the same amount of charge changes significantly as time goes by. Therefore, a comparator that can make an accurate judgment regardless of the varying duration is needed.

ΔΣ Delay Cell Modulation (DSM) Sensor Phase Control



The duration of TRANS and REPDELAYdo not change significantly even after



To solve the problem, the series of two comparators were used. The first one operates in slow speed but consumes low energy. This was realized by a self-clocked dynamic comparator, whose clock speed depends on the voltage level of sensing point. The other operates in fast speed but consumes lots of energy, which was realized by a continuous comparator. Using these two comparators, it was able to make a comparator that works accurately and energy-efficiently.

several repeated cycles. Thus, the $\Delta\Sigma$ delay cell was used which saves the duration time and adjusts it by a charge pump. This operates energy-efficiently, because the comparator judges only once during the phase.

During DELAY_ON, either when it is TRANS or REP phase signal is on, the dependent current source charges the capacitor, a delay cell. When the capacitor voltage exceeds the threshold of the inverter, the DELAY_ON signal is turned off, discharging the capacitor, and moving to the next phase. The charge pump adjusts the current level of dependent current source. According to the information transmitted from the comparator, it increases or decreases V_G. For the fast determination of the duration, digital logic was implemented, which integrates the error information in the flip-flop cell. A degenerative resistor is placed at the source terminal of the dependent current source, increasing the linearity of the current change according to V_G.

		Resu	lt & C	onc	clusion	
Ch	ip Layo	ut				
-		2.5mm		-	2.5mm	
2mm	Sensors Biasir Circu	Tracking Cap. Digital Controllers		2mm	Stimulation Site Tracking Cap. Digital Controllers Biasing Circuit	

Result

The following results are simulated by ADE tool from Cadence. The PCB experiment was failed due to an error in the initial stage of the chip.

Waveform of stimulation current





Charge Tracking Error

$\mathbf{R}_{\mathbf{t}}$	Target Charge	Delivered Charge	Error
500Ω		318.12nC	6.04%
$1\mathrm{k}\Omega$	300nC	316.66nC	5.55%
$5\mathrm{k}\Omega$		$315.28 \mathrm{nC}$	4.85%

Charge Balancing Error

R_t	\mathbf{CS}	AS	Residuals			Er	ror	
			B/F I	Passive	A/F I	Passive	B/F Passive	A/F Passive
			Charge	Voltage	Charge	Voltage		
500Ω	$318.12 \mathrm{nC}$	319.06nC	$944 \mathrm{pC}$	$3.76 \mathrm{mV}$	$4.10 \mathrm{fC}$	$16.4 \mathrm{nV}$	0.29%	0%
$1\mathrm{k}\Omega$	316.66 nC	$314.80 \mathrm{nC}$	$1.86 \mathrm{nC}$	$7.26 \mathrm{mV}$	$86.3 \mathrm{aC}$	$345 \mathrm{pV}$	0.59%	0%
$5\mathrm{k}\Omega$	$315.28 \mathrm{nC}$	$310.95 \mathrm{nC}$	$4.28 \mathrm{nC}$	$17.1 \mathrm{mV}$	$12.2 \mathrm{pC}$	$48.6\mathrm{uV}$	1.36%	0%

The integrated circuit is designed with a TSMC 180nm RF CMOS process. The size is 2.5mm x 2mm, an area of 5 mm². **Electrode-tissue Interference Model**



Rt=500 Ω , 1k Ω , 5k Ω , Rd=10M Ω , and Cd=500nF

Acknowledgement

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Stimulation Efficiency

	R_t	
500Ω	$1 \mathrm{k} \Omega$	$5\mathrm{k}\Omega$
$1.073 \mu J/1.273 \mu J$	1.101µJ/1.280µJ	1.114µJ/1.278µJ
$\rightarrow 84.3\%$	$\rightarrow 86.0\%$	$\rightarrow 87.2\%$

Stimulator Efficiency

	R _t	
500Ω	$1 \mathrm{k} \Omega$	$5k\Omega$
$1.073 \mu J / 1.340 \mu J$	$1.101 \mu J/1.384 \mu J$	1.114µJ/1.665µJ
$\rightarrow 80.1\%$	$\rightarrow 79.6\%$	$\rightarrow 66.9\%$

